



On the Design of Arbitrarily Low-Rate Turbo Codes

Miguel Griot
Andres I. Vila Casado
Richard Wesel

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Low-rate codes: Motivation

■ Practical:

- Applications requiring very low SNRs: deep-space networks.
- Spread-spectrum multiple-access channels: CDMA.
- Interleave-Division Multiple-Access : IDMA [Li Ping'06].
- Rateless coding.

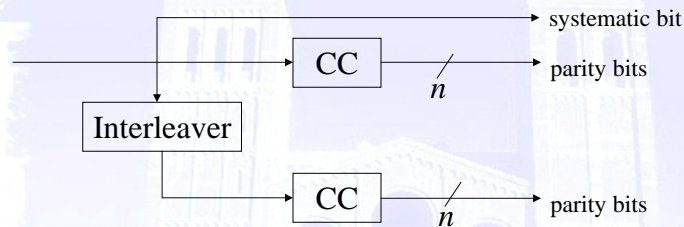
■ Theoretical:

- Approach the ultimate low-rate Shannon limit of $E_b / N_0 \approx -1.59dB$ on the AWGN Channel.

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Low-rate turbo codes

- We consider low-rate turbo codes:



- Given a certain number of trellis states of each constituent encoder:

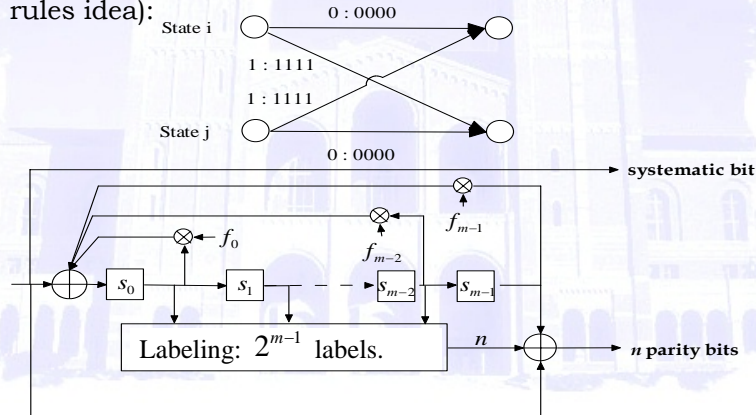
- Goal: approach ultimate Shannon capacity $E_b / N_0 \approx -1.59dB$.
- What is the best rate? Structure?

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Previous work

- Super-orthogonal turbo codes:

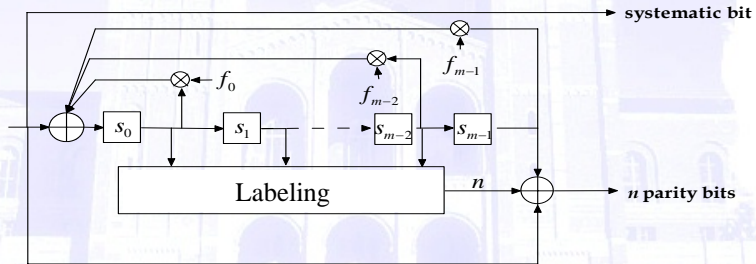
- For a certain number of states of the constituent code 2^m and a certain trellis structure.
- Use antipodal labeling on splits and merges (Ungerboeck's rules idea):



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Questions

- To optimize performance vs. E_b/N_0 :
 - Given a certain rate and number of states of the constituent encoder, what is the best structure and labeling?
 - What is the best set of rates?

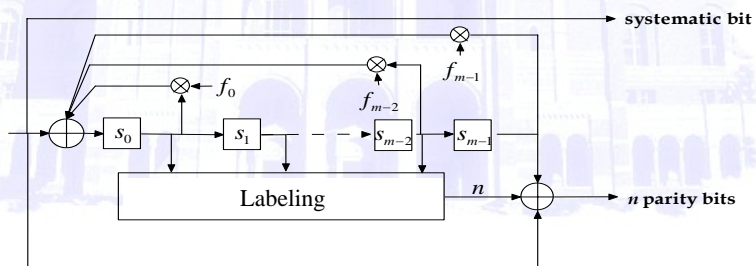


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General Constituent Encoder Structure

- Given a certain number of states 2^m then the recursive structure of the trellis can be chosen to maximize the length of trellis error events with input Hamming distance 2 (effective free distance). This number is 2^m .

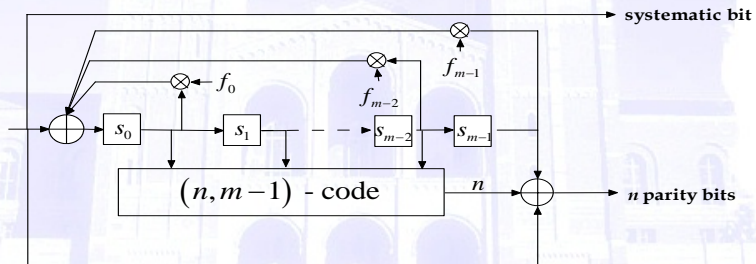
m	$f_0 f_1 \dots f_{m-1}$
2	11
3	011
4	0011
5	00101



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Labeling

- Use antipodal output labeling in splits and merges.
- The output labeling design can be seen as a code design.



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What is the optimal value of n ?

- The larger n the larger the minimum distance of the code for labeling.
- We want to optimize the code in terms of BER vs. E_b / N_0
- Now for an (n, k) code and an AWGN channel:

$$P_e \approx Q\left(\sqrt{2d_{\min} E_s / N_0}\right) = Q\left(2(k/n)d_{\min} E_b / N_0\right)$$

- $k = m-1$ fixed.
- Criteria: maximize $\frac{d_{\min}}{n}$
- Although the rate of the turbo code is $1/(2n+1)$, the effect of the systematic bit has little effect on the performance of low-rate turbo codes.

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BCH codes are optimal

- Theorem I: every (n,k) code must satisfy:

$$\frac{d_{\min}}{n} \leq \frac{2^{k-1}}{2^k - 1}$$

- Proof for linear codes: linear codes produce equally likely ones and zeros. Consider the sum of the distances between the all-zero codeword and non-zero codewords.

$$(2^k - 1)d_{\min} \leq \sum_{i=1}^{2^k-1} d_H(0, c_i)$$

$$\sum_{i=1}^{2^k-1} d_H(0, c_i) \leq \sum_{i=0}^{2^k-1} W_H(c_i) = 2^k \binom{n}{2} = 2^{k-1}n$$

- For nonlinear codes with unequally likely ones and zeros, strict inequality.

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BCH codes are optimal

- Theorem I: every (n,k) code must satisfy:

$$\frac{d_{\min}}{n} \leq \frac{2^{k-1}}{2^k - 1}$$

- Theorem II: Equality holds when $n = 2^k - 1$ using a BCH code or when $n = r(2^k - 1)$ using a BCH code with an $(r,1)$ repetition code.

- In that case $d_H(c_i, c_j) = d_{\min} = r \times 2^{k-1}$.

- This provides the labeling and the rate.

- Rate =
$$\frac{1}{1 + 2r(2^{m-1} - 1)}$$

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BCH codes

- BCH codes can be represented with generator polynomials. In this case, all $2^{m-2} - 1$ possible polynomials of the form:

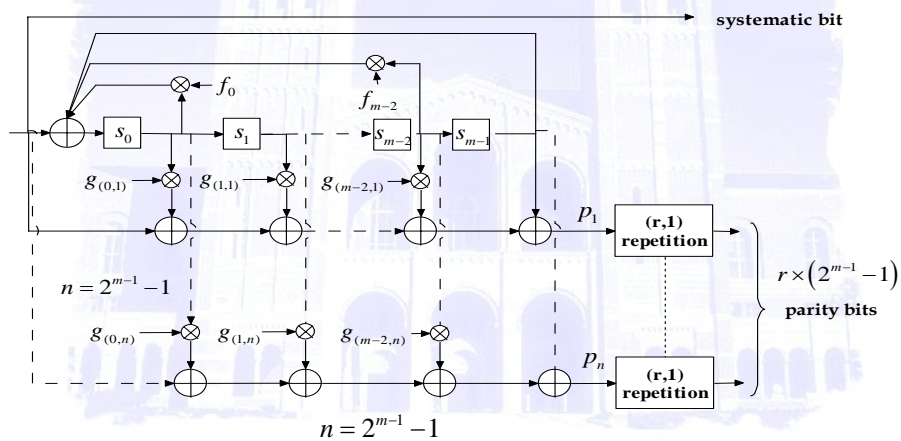
$$g(x) = a_0 + a_1x + \dots + a_{m-2}x^{m-2} \neq 0$$

- They can also be represented by the rows of a Hadamard matrix.

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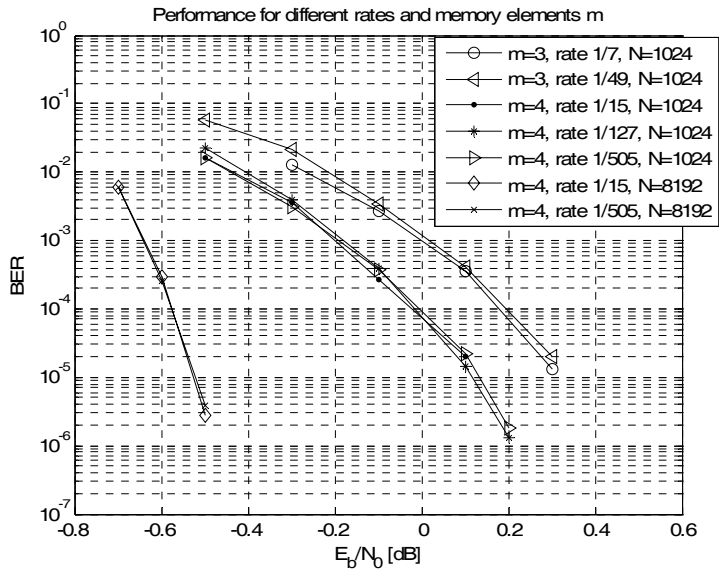
Labeling of Linear constituent encoder

- BCH codes can be represented with generator polynomials.



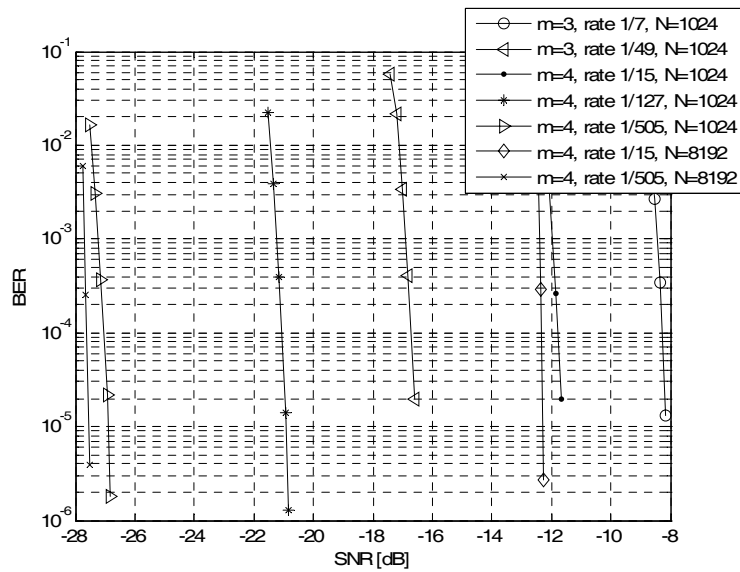
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Performance: Eb/No



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Performance: SNR



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Conclusions

- We have shown a criteria to design low-rate turbo codes.
- Goal: Minimize BER vs. Eb/No.
- We see the output labeling problem as a coding problem.
- Criteria: maximize $\frac{d_{\min}}{n}$
 - Given the number of states 2^m , provides set of rates as well as structure and labeling,
 - $(2^{m-1}-1, m-1)$ BCH codes are optimal for labeling under this criteria.
 - $n = r(2^{m-1}-1)$